







Use of simple linear regression for transfer function model learning – A brief recap

When the model structure is known

- In the ideal noise-free or deterministic case, use of simple least squares works fine !
 - the transfer function model parameter can be estimated by linear regression
- In practice, the simple least squares method breaks down
 - The output is not perfectly known. It is contaminated by measurement noise

 \Rightarrow Incorrect least squares estimates (whatever the continuous or discrete-time model form)



How can we model the noisy measurement output ?







Identifying measurement noise models

- In control design and estimation/prediction, it is often important to identify not only the dynamics from input to output, but also the measurement noise dynamics
 - how noise and disturbance perturb the system
 - where noise comes in
 - whether the noise is colored and correlated
- In system identification, various techniques are available for identifying both input-output dynamics and noise dynamics





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Models for disturbance and measurement noise

So far: only deterministic models



 $u(t) = \mathsf{input}$

- w(t) = disturbance
- e(t) = measurement noise

$$y(t) =$$
 output

- disturbance and measurement noise are stochastic signals
- Stochastic models:
 - means, covariances
 - spectra (energy or power)





Moments of a probability distribution A brief review

- Moments of a random variable X with density $f_X(x)$:
 - *l*-th moment

$$m'_l = \mathbb{E}[X^l] = \int_{-\infty}^{\infty} x^l f_X(x) \, dx$$

- *l*-th central moment

$$m_l = \mathbb{E}[(X-\mu)^l] = \int_{-\infty}^{\infty} (x-\mu)^l f_X(x) \, dx$$

- Low-order moments: example
 - Expectation (mean): $m_1 = \mu = E[X]$
 - Variance: $m_2 = \sigma^2 = E[(X \mu)^2]$





- AutoCorrelation Function (ACF)





Major assumption: **stationarity** of the signals

- The properties of one section of a data are much like the properties of the other sections. The future is "similar" to the past (*in a probabilistic sense*)
- A stationary stochastic signals has
 - no trend / no seasonality

- no systematic change in variation

- no periodic fluctuations









Autocorrelation function (ACF)

- Statistical correlation summarizes the strength of the relationship between two different variables
- We can calculate the correlation for time series observations with observations with previous time instants, called lags. This is called an autocorrelation
- A plot of the autocorrelation of a time series in terms of lags is called the AutoCorrelation Function, or its acronym ACF
- Sample ACF at lag h, denoted as $\gamma_y(h)$, measures the linear correlation between y_t and y_{t+h}









Autocorrelation function (ACF)

- ACF: measures the speed of variation of temporal evolutions
 - we compare the time series with itself but shifted by $\tau(or h)$
 - it allows us to see how the time series at a given time is influenced (*linear autocorrelation*) by what happened at a previous time

fluctuations lentes : forte dépendance entre valeurs fluctuations rapides: faible dépendance entre valeurs Successives





Autocorrelation function (ACF)

Slowly varying autocorrelation function – slowly varying process Quickly varying autocorrelation function – quickly varying process







Finite sample statistics

- Given $\{y_1, \dots, y_N\}$ observations of a stationary signal $\{y_t\}$, estimate the finite sample mean, variance, autocovariance and ACF
 - Sample mean

$$\hat{\mu} = \bar{y} = \frac{1}{N} \sum_{i=1}^{N} y_i$$

- Sample variance

$$\hat{\sigma}^2 = \frac{1}{N-1} \sum_{i=1}^{N} (y_i - \hat{\mu})^2$$

Sample autocovariance function

$$\hat{\gamma}_{y}(h) = \frac{1}{N} \sum_{j=1}^{N-h} (y_{j+h} - \bar{y}) (y_{j} - \bar{y}), \quad 0 \le h < N,$$

with
$$\hat{\gamma}_y(h) = \hat{\gamma}_y(-h), \qquad -N < h \le 0$$

Sample autocorrelation function (ACF)

$$\hat{\rho}_{y}(h) = \frac{\hat{\gamma}_{y}(h)}{\hat{\gamma}_{y}(0)}, \qquad |h| < N$$





Finite sample ACF - Example

$$y = [0 \ 1 \ 1 \ 1 \ 0] \qquad N=5$$

$$\bar{y} = \frac{1}{5} \sum_{i=1}^{5} y_i = 0.6$$

$$\hat{y}_y(h) = \frac{1}{5} \sum_{j=1}^{5-h} (y_{j+h} - \bar{y})(y_j - \bar{y}), \qquad h = 0, 1, 2, 3, 4$$

$$\hat{\rho}_{y}(h) = \frac{\hat{\gamma}_{y}(h)}{\hat{\gamma}_{y}(0)}, \qquad h = 0, 1, 2, 3, 4$$

$$\hat{\rho}_y = [1 - 0.13 - 0.26 - 0.4 \ 0.3]$$

In Matlab :

y=[0 1 1 1 0]; [rho_hat_y,Lag]=xcov(y,'norm'); stem(Lag,rho_hat_y) Or

autocorr(y)

$$\hat{\gamma}_{y}(0) = \frac{1}{5} \sum_{j=1}^{5} (y_{j} - \bar{y})(y_{j} - \bar{y}) = 0.24$$
$$\hat{\gamma}_{y}(1) = \frac{1}{5} \sum_{j=1}^{4} (y_{j+1} - \bar{y})(y_{j} - \bar{y}) = -0.0320$$
$$\hat{\gamma}_{y}(2) = \frac{1}{5} \sum_{j=1}^{3} (y_{j+2} - \bar{y})(y_{j} - \bar{y}) = -0.0620$$
$$\hat{\gamma}_{y}(3) = \frac{1}{5} \sum_{j=1}^{2} (y_{j+3} - \bar{y})(y_{j} - \bar{y}) = -0.0960$$
$$\hat{\gamma}_{y}(4) = \frac{1}{5} \sum_{j=1}^{1} (y_{j+4} - \bar{y})(y_{j} - \bar{y}) = 0.0720$$







The white noise process The most fundamental example of stationary signal

- A white noise is a sequence of independent and identically distributed (i.i.d) random variables
 - The sequences are *uncorrelated*, have zero mean, and constant variance
 - A Gaussian white noise are *i.i.d* observations from $\mathcal{N}(0, \sigma^2)$
 - Because independence implies that its variables are uncorrelated at different times, its ACF looks like a Kronecker impulse









Finite sample distribution of sample ACF

- Finite sample distribution of ACF for a white noise is asymptotically Gaussian $\mathcal{N}\left(0,\frac{1}{N}\right)$
 - 95% of all ACF coefficients for a white noise must lie within $\pm \frac{1.96}{\sqrt{N}}$
 - It is common to plot horizontal limit lines at $\pm \frac{1.96}{\sqrt{N}}$ when plotting the ACF
- When N = 125, limit line values at $\pm \frac{1.96}{\sqrt{125}} = \pm 0.175$
 - All ACF coefficients lie within these limits, confirming that the data are white noise (more precisely, the data cannot be distinguished from white noise)







Properties of white noise

- Best prediction of a white noise
 - If a signal is white noise, it is unpredictable and so there is nothing to forecast.
 Or more precisely, the best prediction is its mean value which is zero
- Whitening test of the residuals
 - At the validation stage of the system identification methodology, we will check whether the prediction errors=residuals are a white noise by plotting its sample ACF



- Sample ACF shows some significant autocorrelations at lags 1, 2, 3 and 4. This shows the residuals are not white here
- If the residual ACF does not resemble to the ACF of a white noise, it suggests that improvements could be made to the predictive model
- If the residual ACF resembles to the ACF of a white noise, the modelling procedure is finished. There is nothing else to capture in the residuals













General linear parametric model of stationary signals

- Box and Jenkins in 1970 (following Yule and Slutsky 1927)
 - Many time series (or their derivatives) can be considered as a special class of stochastic processes: (weakly) stationary stochastic processes
 - First two moments are finite and constant over time
 - Defined completely by the mean, variance and autocorrelation function
- General parametric model of stationary stochastic processes (Wold 1938)
 - All (weakly) stationary stochastic processes can be written as

$$y_k = c + \sum_{i=1}^{+\infty} \psi_i e_{k-i} + e_k$$

where c is a constant and e_k is a white Gaussian noise

- e_k is often called the *innovation process* because it captures all new information in the series at time k





Backward shift q^{-1} operator

• The **backward shift** operator, q^{-1} , is defined as

$$q^{-1} \varepsilon_k = \varepsilon_{t-1}$$

$$q^{-i} \varepsilon_k = \varepsilon_{k-i}$$

• The general linear model of a stationary stochastic process can be written as

 $y_{k} = c + \sum_{i=1}^{+\infty} \psi_{i} \varepsilon_{k-i} + \varepsilon_{k}$ $y_{k} = c + \psi(q^{-1}) \varepsilon_{k}$

$$\Psi(q^{-1}) = 1 + \sum_{i=1}^{+\infty} \psi_i L^i$$

• This model has an infinite-degree polynomial $\Psi(q^{-1})$ with infinite coefficients which cannot be estimated from a finite amount of data in the time series \cong





Towards AR, MA and ARMA models for stationary signals

• If *H*(*q*) is a rational polynomial, we can write it (at least approximately) as the quotient of two finite-degree polynomials

$$H(q) = \frac{C(q^{-1})}{D(q^{-1})}$$

$$C(q^{-1}) = 1 + c_1 q^{-1} + \dots + c_{n_c} q^{-n_c}$$

$$D(q^{-1}) = 1 + d_1 q^{-1} + \dots + d_{n_d} q^{-n_d}$$

(Matlab System Identification toolbox notations)

• Wold's theorem: every zero-mean stationary stochastic process can be written as

$$y_k = \frac{\mathcal{C}(q^{-1})}{\mathcal{D}(q^{-1})} e_k$$

- which has a finite number $(n_c + n_d)$ of coefficients
- e_k is a white Gaussian noise
- This leads to the use of <u>parsimonious</u> models : **AR**, **MA** and **ARMA** models
 - They are most useful for practical applications since these models can be quite easily estimated from a finite number of signal samples

