# Design of fault tolerant control for nonlinear systems subject to time varying faults

Tahar Bouarar, Benoît Marx, Didier Maquin, José Ragot

Centre de Recherche en Automatique de Nancy (CRAN) Nancy, France

ESREL, September 18-22, 2011, Troyes France







#### Motivations



### Objective of diagnosis and fault tolerant control

- ► To detect, isolate and estimate the actuator fault (diagnosis)
- ▶ To modify the control law to accommodate the fault (FTC)

#### Motivations



### Objective of diagnosis and fault tolerant control

- To detect, isolate and estimate the actuator fault (diagnosis)
- To modify the control law to accommodate the fault (FTC)

#### **Difficulties**

- Taking into account the system complexity in a large operating range
- Nonlinear behavior of the system
- The faults are time varying

#### **Motivations**



### Objective of diagnosis and fault tolerant control

- To detect, isolate and estimate the actuator fault (diagnosis)
- To modify the control law to accommodate the fault (FTC)

#### **Difficulties**

- Taking into account the system complexity in a large operating range
- Nonlinear behavior of the system
- The faults are time varying

### Proposed strategy

- Takagi-Sugeno representation of nonlinear systems
- Extension of the existing results on linear systems
- Observer-based fault tolerant control design
- Consideration of an a priori model of the fault

#### Outline

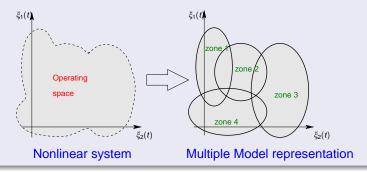


- Takagi-Sugeno approach for modeling
- Observer and FTC law structures
- A priori considered fault models
- Controller design
- Simulations results
- 6 Conclusions

### Multiple models principle



- Operating range decomposition in several local zones.
- ▶ A simple submodel represents the behavior of the system in a specific zone.
- The overall behavior of the system is obtained by aggregating the submodels with adequate weighting functions.





### The main idea of Takagi-Sugeno approach

- ▶ Define local models  $M_i$ , i = 1..r
- ▶ Define weighting functions  $\mu_i(\xi)$ , s.t.  $0 \le \mu_i \le 1$  and  $\sum_{i=1}^r \mu_i(\xi) = 1$
- $\rightarrow$  the global model is obtained by aggregation :  $M = \sum_{i=1}^{r} \mu_i(\xi) M_i$



### The main idea of Takagi-Sugeno approach

- ▶ Define local models  $M_i$ , i = 1..r
- ▶ Define weighting functions  $\mu_i(\xi)$ , s.t.  $0 \le \mu_i \le 1$  and  $\sum_{i=1}^r \mu_i(\xi) = 1$
- $\rightarrow$  the global model is obtained by aggregation :  $M = \sum_{i=1}^{r} \mu_i(\xi) M_i$

### Interests of Takagi-Sugeno approach

- The specific study of the nonlinearities is not required.
- Analysis (stability, performance, robustness, etc.) and design (controller, observer, etc.) are based on the linear submodels.
- → Possible extension of the theoretical LTI tools for nonlinear systems.



#### The main idea of Takagi-Sugeno approach

- ▶ Define local models  $M_i$ , i = 1..r
- ▶ Define weighting functions  $\mu_i(\xi)$ , s.t.  $0 \le \mu_i \le 1$  and  $\sum_{i=1}^r \mu_i(\xi) = 1$
- $\rightarrow$  the global model is obtained by aggregation :  $M = \sum_{i=1}^{r} \mu_i(\xi) M_i$

### Interests of Takagi-Sugeno approach

- The specific study of the nonlinearities is not required.
- Analysis (stability, performance, robustness, etc.) and design (controller, observer, etc.) are based on the linear submodels.
- → Possible extension of the theoretical LTI tools for nonlinear systems.

#### The difficulties

- How many local models?
- How to define the domain of influence of each local model?
- On what variables may depend the weighting functions μ<sub>i</sub>?



### Obtaining a Takagi-Sugeno model

- Identification approach
  - Choice of premise variables
  - Choice of the structure of the local models
  - Parameter identification
- Transformation of an a priori known nonlinear model
  - Linearization around some points
    - how to chose the linearization points?
    - how to define the weighting functions, minimizing the approximation error
  - Nonlinear sector approach

Equivalent rewriting of the model in a compact set of the state space

$$\begin{cases} x(k+1) = f(x(k), u(k)) \\ y(k) = h(x(k), u(k)) \end{cases} \Rightarrow \begin{cases} x(k+1) = \sum_{i=1}^{r} \mu_i(\xi(k)) (A_i x(k) + B_i u(k)) \\ y(k) = \sum_{i=1}^{r} \mu_i(\xi(k)) (C_i x(k) + D_i u(k)) \end{cases}$$



### Obtaining a Takagi-Sugeno model

- Identification approach
  - Choice of premise variables
  - Choice of the structure of the local models
  - Parameter identification
- Transformation of an a priori known nonlinear model
  - Linearization around some points
    - how to chose the linearization points?
    - how to define the weighting functions, minimizing the approximation error
    - Nonlinear sector approach

Equivalent rewriting of the model in a compact set of the state space

$$\begin{cases} x(k+1) = f(x(k), u(k)) \\ y(k) = h(x(k), u(k)) \end{cases} \Rightarrow \begin{cases} x(k+1) = \sum_{i=1}^{r} \mu_i(\xi(k)) (A_i x(k) + B_i u(k)) \\ y(k) = \sum_{i=1}^{r} \mu_i(\xi(k)) (C_i x(k) + D_i u(k)) \end{cases}$$



### Obtaining a Takagi-Sugeno model

- Identification approach
  - Choice of premise variables
  - Choice of the structure of the local models
  - Parameter identification
- Transformation of an a priori known nonlinear model
  - Linearization around some points
    - how to chose the linearization points?
    - how to define the weighting functions, minimizing the approximation error
  - Nonlinear sector approach

Equivalent rewriting of the model in a compact set of the state space

$$\begin{cases} x(k+1) = f(x(k), u(k)) \\ y(k) = h(x(k), u(k)) \end{cases} \Rightarrow \begin{cases} x(k+1) = \sum_{i=1}^{r} \mu_i(\xi(k)) (A_i x(k) + B_i u(k)) \\ y(k) = \sum_{i=1}^{r} \mu_i(\xi(k)) (C_i x(k) + D_i u(k)) \end{cases}$$

# Takagi-Sugeno system



#### Reference model

$$\begin{cases} x(k+1) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) (A_{i}x(k) + B_{i}u(k)) \\ y(k) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) (C_{i}x(k) + D_{i}u(k)) \end{cases}$$

- Interpolation mechanism  $\sum\limits_{i=1}^r \mu_i(\xi(k)) = 1$  and  $0 \le \mu_i(\xi(k)) \le 1, \forall k, \forall i \in \{1,...,r\}$
- The premise variable  $\xi(k)$  are measurable (like u(k), y(k)).



#### Reference model

$$\begin{cases} x(k+1) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) (A_{i}x(k) + B_{i}u(k)) \\ y(k) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) (C_{i}x(k) + D_{i}u(k)) \end{cases}$$

- Interpolation mechanism  $\sum\limits_{i=1}^r \mu_i(\xi(k)) = 1$  and  $0 \le \mu_i(\xi(k)) \le 1, \forall k, \forall i \in \{1,...,r\}$
- The premise variable  $\xi(k)$  are measurable (like u(k), y(k)).

### The faulty system

$$\begin{cases} x_f(k+1) = \sum_{i=1}^r \mu_i(\xi(k)) (A_i x_f(k) + B_i u_f(k) + G_i f(k)) \\ y_f(k) = \sum_{i=1}^r \mu_i(\xi(k)) (C_i x_f(k) + D_i u_f(k) + W_i f(k)) \end{cases}$$

• f(k) represents the fault vector to be detected and accommodated.

### Fault tolerant control design



### Objectives: estimation + diagnosis + FTC

- estimate the faulty system state  $x_f(k)$
- estimate the occurring fault f(k)
- ▶ reconfigure the control law for trajectory tracking  $x_f(k) \rightarrow x(k)$

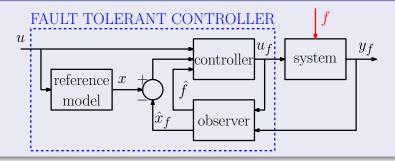
### Fault tolerant control design



#### Objectives: estimation + diagnosis + FTC

- estimate the faulty system state  $x_f(k)$
- estimate the occurring fault f(k)
- ▶ reconfigure the control law for trajectory tracking  $x_f(k) \rightarrow x(k)$

#### Fault tolerant control scheme



#### Observer and FTC law structures



#### Faulty system

$$\begin{cases} x_f(k+1) = \sum_{i=1}^r \mu_i(\xi(k)) (A_i x_f(k) + B_i u_f(k) + G_i f(k)) \\ y_f(k) = \sum_{i=1}^r \mu_i(\xi(k)) (C_i x_f(k) + D_i u_f(k) + W_i f(k)) \end{cases}$$

#### Observer and FTC law structures



#### Faulty system

$$\begin{cases} x_f(k+1) = \sum_{i=1}^r \mu_i(\xi(k)) (A_i x_f(k) + B_i u_f(k) + G_i f(k)) \\ y_f(k) = \sum_{i=1}^r \mu_i(\xi(k)) (C_i x_f(k) + D_i u_f(k) + W_i f(k)) \end{cases}$$

#### PI Observer

$$\begin{cases} \hat{x}_{f}(k+1) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) \left( A_{i}\hat{x}_{f}(k) + B_{i}u_{f}(k) + G_{i}\hat{f}(k) + H_{i}^{1}(y_{f}(k) - \hat{y}_{f}(k)) \right) \\ \hat{f}(k+1) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) \left( H_{i}^{2}(y_{f}(k) - \hat{y}_{f}(k)) + \hat{f}(k) \right) \\ \hat{y}_{f}(k) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) \left( C_{i}\hat{x}_{f}(k) + D_{i}u_{f}(k) + W_{i}\hat{f}(k) \right) \end{cases}$$

### Observer and FTC law structures



#### Faulty system

$$\begin{cases} x_f(k+1) = \sum_{i=1}^r \mu_i(\xi(k)) (A_i x_f(k) + B_i u_f(k) + G_i f(k)) \\ y_f(k) = \sum_{i=1}^r \mu_i(\xi(k)) (C_i x_f(k) + D_i u_f(k) + W_i f(k)) \end{cases}$$

#### PI Observer

$$\begin{cases} \hat{x}_{f}(k+1) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) \left( A_{i} \hat{x}_{f}(k) + B_{i} u_{f}(k) + G_{i} \hat{f}(k) + H_{i}^{1}(y_{f}(k) - \hat{y}_{f}(k)) \right) \\ \hat{f}(k+1) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) \left( H_{i}^{2}(y_{f}(k) - \hat{y}_{f}(k)) + \hat{f}(k) \right) \\ \hat{y}_{f}(k) = \sum_{i=1}^{r} \mu_{i}(\xi(k)) \left( C_{i} \hat{x}_{f}(k) + D_{i} u_{f}(k) + W_{i} \hat{f}(k) \right) \end{cases}$$

#### FTC law

$$u_f(k) = u(k) + \sum_{i=1}^r \mu_i(\xi(k)) \left( \frac{K_i(x(k) - \hat{x}_f(k)) - \hat{f}(k)}{k} \right)$$



### **Exponential faults**

$$f_i(k) = \mathbf{e}^{lpha_i k + eta_i}, ext{ with } lpha_i, eta_i \in \mathbb{R}, i = 1,...,q$$
  $lpha_i = lpha_{0,i} + \Delta lpha_i$ 

where  $\alpha_{0,i}$  and  $\Delta \alpha_i$  are respectively the nominal and the uncertain parts of  $\alpha_i$ 

Let us define :

$$egin{aligned} &lpha = \operatorname{diag}(lpha_1,...,lpha_q) \ &lpha_0 = \operatorname{diag}(lpha_{0,1},...,lpha_{0,q}) \ &\Deltalpha = \operatorname{diag}(\Deltalpha_1,...,\Deltalpha_q) \end{aligned}$$

The uncertain part can be bounded as:

$$(\Delta \alpha)^T \Delta \alpha \leq \lambda$$

where  $\lambda \in \mathbb{R}^{q \times q}$  is a known diagonal positive definite matrix.





#### **Estimation errors**

$$\left\{ \begin{array}{l} e_p(k) = x(k) - x_f(k) \ : \ \text{state tracking error} \\ e_s(k) = x_f(k) - \hat{x}_f(k) \ : \ \text{state estimation error} \\ e_d(k) = f(k) - \hat{f}(k) \ : \ \text{fault estimation error} \end{array} \right.$$

# Controller design – the exponential fault case



#### **Estimation errors**

$$\left\{ \begin{array}{l} e_p(k) = x(k) - x_f(k) \ : \ \text{state tracking error} \\ e_s(k) = x_f(k) - \hat{x}_f(k) \ : \ \text{state estimation error} \\ e_d(k) = f(k) - \hat{f}(k) \ : \ \text{fault estimation error} \end{array} \right.$$

### Notation and hypothesis

$$X_{\mu} = \sum_{i=1}^{r} \mu_{i}(\xi(k))X_{i} \qquad X_{\mu\mu} = \sum_{i=1}^{r} \sum_{j=1}^{r} \mu_{i}(\xi(k))\mu_{j}(\xi(k))X_{ij} \qquad f_{i}(k+1) = e^{\alpha_{i}}f_{i}(k)$$

# Controller design - the exponential fault case



#### **Estimation errors**

$$\left\{ \begin{array}{l} e_p(k) = x(k) - x_f(k) \ : \ \text{state tracking error} \\ e_s(k) = x_f(k) - \hat{x}_f(k) \ : \ \text{state estimation error} \\ e_d(k) = f(k) - \hat{f}(k) \ : \ \text{fault estimation error} \end{array} \right.$$

#### Notation and hypothesis

$$X_{\mu} = \sum_{i=1}^{r} \mu_{i}(\xi(k))X_{i}$$
  $X_{\mu\mu} = \sum_{i=1}^{r} \sum_{j=1}^{r} \mu_{i}(\xi(k))\mu_{j}(\xi(k))X_{ij}$   $f_{i}(k+1) = e^{\alpha_{i}}f_{i}(k)$ 

### Dynamics of the tracking and estimation errors

$$\underbrace{\begin{pmatrix} e_p(k+1) \\ e_s(k+1) \\ e_d(k+1) \end{pmatrix}}_{\overline{e}(k+1)} = \underbrace{\begin{pmatrix} A_{\mu\mu} - B_{\mu} \textbf{K}_{\mu} & -B_{\mu} \textbf{K}_{\mu} & -B_{\mu} \\ 0 & A_{\mu} - \textbf{H}_{\mu}^1 C_{\mu} & G_{\mu} - \textbf{H}_{\mu}^1 \textbf{W}_{\mu} \\ 0 & -\textbf{H}_{\mu}^2 C_{\mu} & I - \textbf{H}_{\mu}^2 \textbf{W}_{\mu} \end{pmatrix}}_{\overline{e}(k)} \underbrace{\begin{pmatrix} e_p(k) \\ e_s(k) \\ e_d(k) \end{pmatrix}}_{\overline{e}(k)} + \underbrace{\begin{pmatrix} B_{\mu} - G_{\mu} \\ 0 \\ \alpha - I \end{pmatrix}}_{\overline{B}_{\mu}} f(k)$$

# Controller design - the exponential fault case.



The tracking, state estimation and fault estimation errors are ruled by :

$$\overline{e}(k+1) = \overline{A}_{\mu\mu}\overline{e}(k) + \overline{B}_{\mu}f(k)$$

The FTC design reduces to find the controller and observer gains :  $K_i$ ,  $H_i^1$  and  $H_i^2$  satisfying the two main objectives.

### Tracking, state and fault estimation error convergence in the fault free case

Find a positive definite Lyapunov function such that

$$\Delta V(k) = V(k+1) - V(k) < 0$$

Here, a quadratic Lyapunov function is chosen:

$$V(k) = \overline{e}^T(k)X\overline{e}(k)$$
, with  $X = X^T > 0$ 

# Controller design – the exponential fault case



The tracking, state estimation and fault estimation errors are ruled by :

$$\overline{e}(k+1) = \overline{A}_{\mu\mu}\overline{e}(k) + \overline{B}_{\mu}f(k)$$

The FTC design reduces to find the controller and observer gains :  $K_i$ ,  $H_i^1$  and  $H_i^2$  satisfying the two main objectives.

### Tracking, state and fault estimation error convergence in the fault free case

Find a positive definite Lyapunov function such that

$$\Delta V(k) = V(k+1) - V(k) < 0$$

Here, a quadratic Lyapunov function is chosen:

$$V(k) = \overline{e}^T(k)X\overline{e}(k)$$
, with  $X = X^T > 0$ 

#### Attenuation of the fault effect

The  $\mathscr{L}_2$ -gain from the fault f(k) to the errors  $\overline{e}(k)$  is bounded by a positive  $\gamma$ 

$$\sum_{k=1}^{N} \overline{\mathbf{e}}^{T}(k) \mathbf{Q} \overline{\mathbf{e}}(k) \leq \gamma^{2} \sum_{k=1}^{N} f^{T}(k) f(k)$$

# Controller design - the exponential fault case



### Summary

The tracking error  $e_p(k)$ , state and fault estimation errors  $e_s(k)$  and  $e_d(k)$  must therefore satisfy the following inequality :

$$\overline{e}^T(k+1)X\overline{e}(k+1) - \overline{e}^T(k)X\overline{e}(k) + \overline{e}^T(k)Q\overline{e}(k) - \gamma^2 f^T(k)f(k) < 0$$

This inequality is fulfilled if:

$$\begin{pmatrix} Q - X & 0 \\ 0 & -\gamma^2 I \end{pmatrix} + \begin{pmatrix} \overline{A}_{\mu\mu}^T \\ \overline{B}_{\mu}^T \end{pmatrix} X \begin{pmatrix} \overline{A}_{\mu\mu} & \overline{B}_{\mu} \end{pmatrix} < 0$$

# Controller design - the exponential fault case \_



### Summary

The tracking error  $e_p(k)$ , state and fault estimation errors  $e_s(k)$  and  $e_d(k)$  must therefore satisfy the following inequality :

$$\overline{e}^T(k+1)X\overline{e}(k+1) - \overline{e}^T(k)X\overline{e}(k) + \overline{e}^T(k)Q\overline{e}(k) - \gamma^2 f^T(k)f(k) < 0$$

This inequality is fulfilled if:

$$\begin{pmatrix} Q - X & 0 \\ 0 & -\gamma^2 I \end{pmatrix} + \begin{pmatrix} \overline{A}_{\mu\mu}^T \\ \overline{B}_{\mu}^T \end{pmatrix} X \begin{pmatrix} \overline{A}_{\mu\mu} & \overline{B}_{\mu} \end{pmatrix} < 0$$

- ► Chosing the Lyapunov matrix structure :  $X = \begin{pmatrix} X_1 & 0 & 0 \\ 0 & X_2 & 0 \\ 0 & 0 & X_3 \end{pmatrix}$
- ▶ knowing that  $\mu_i(\xi(k)) \ge 0$
- with some matrix manipulations (Schur complement, S-procedure)
- → sufficient LMI conditions are derived





#### Theorem 1

The tracking and estimation errors asymptotically converge to zero in the fault free case and the  $\mathcal{L}_2$ -gain from f to  $\overline{e}$  is bounded by  $\gamma$ , if there exists matrices  $X_1 \geq 0$ ,  $X_2 \geq 0$ ,  $X_3 \geq 0$ ,  $K_i$ ,  $L_i^1$  and  $L_i^2$  and scalars  $\overline{\gamma}$  and  $\tau$  such that, for i=1,2,...,r

$$\begin{pmatrix} \textbf{Q}_1 - \textbf{X}_1 & 0 & 0 & 0 & * & 0 & 0 & 0 & * & 0 \\ 0 & \textbf{Q}_2 - \textbf{X}_2 & 0 & 0 & 0 & * & * & 0 & 0 & 0 & * \\ 0 & 0 & \textbf{Q}_3 - \textbf{X}_3 & 0 & * & * & * & 0 & 0 & 0 & * \\ 0 & 0 & 0 & \textbf{T}^{-1} \lambda - \overline{\gamma} I & * & 0 & * & 0 & * & 0 & 0 \\ \textbf{X}_1 A_i & 0 & - \textbf{X}_1 B_i & \textbf{X}_1 (B_i - G_i) & - \textbf{X}_1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & \textbf{X}_2 A_i - L_j^1 C_i & \textbf{X}_2 G_i - L_j^1 W_i & 0 & 0 & - \textbf{X}_2 & 0 & 0 & 0 & 0 & 0 \\ 0 & \textbf{L}_j^2 C_i & \textbf{X}_3 - L_j^2 W_i & - \textbf{X}_3 & 0 & 0 & - \textbf{X}_3 & * & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & - \textbf{X}_3 & - \textbf{T}^{-1} I & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \textbf{X}_1 & 0 & 0 & 0 & -2 I & 0 & 0 \\ B_i \textbf{K}_j & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & -I & 0 \\ 0 & B_i \textbf{K}_j & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & -I & 0 \\ \end{pmatrix}$$

The observer gains and the attenuation level are obtained by:

$$H_i^1 = X_2^{-1} L_i^1$$
,  $H_i^2 = X_3^{-1} L_i^2$  and  $\gamma = \sqrt{\bar{\gamma}}$ 

# Controller design – the polynomial fault case



# Polynomial faults

$$f_i(k) = a_i k + b_i$$
, with  $a_i, b_i \in \mathbb{R}, i = 1, ..., q$ 

As well as for exponential function, defining different diagonal matrices,  $a=a_0+\Delta a$ , with  $\Delta a$  verifying :

$$(\Delta a)^T \Delta a \leq \delta$$

where  $\delta \in \mathbb{R}^{q \times q}$  is a known diagonal positive definite matrix.

# Controller design – the polynomial fault case



### Polynomial faults

$$f_i(k) = a_i k + b_i$$
, with  $a_i, b_i \in \mathbb{R}, i = 1, ..., q$ 

As well as for exponential function, defining different diagonal matrices,  $a=a_0+\Delta a$ , with  $\Delta a$  verifying :

$$(\Delta a)^T \Delta a \leq \delta$$

where  $\delta \in \mathbb{R}^{q \times q}$  is a known diagonal positive definite matrix.

### Dynamics of the tracking and estimation errors

Defining  $\overline{e}^T(k) = [e_p^T(k) \ e_s^T(k) \ e_d^T(k)]$ , it follows

$$\overline{e}(k+1) = \underbrace{\begin{pmatrix} A_{\mu\mu} - B_{\mu} K_{\mu} & -B_{\mu} K_{\mu} & -B_{\mu} \\ 0 & A_{\mu} - H_{\mu}^{1} C_{\mu} & G_{\mu} - H_{\mu}^{1} W_{\mu} \\ 0 & -H_{\mu}^{2} C_{\mu} & I - H_{\mu}^{2} W_{\mu} \end{pmatrix}}_{\overline{e}(k)} \underbrace{\begin{pmatrix} e_{p}(k) \\ e_{s}(k) \\ e_{d}(k) \end{pmatrix}}_{\overline{e}(k)} + \underbrace{\begin{pmatrix} B_{\mu} - G_{\mu} \\ 0 \\ 0 \end{pmatrix}}_{\overline{E}_{\mu}} f(k) + \begin{pmatrix} 0 \\ 0 \\ a \end{pmatrix}$$

### Controller design – the polynomial fault case



#### Theorem 2

The tracking and estimation errors asymptotically converge to zero in the fault free case and the  $\mathcal{L}_2$ -gain from f to  $\overline{e}$  is bounded by  $\gamma$ , if there exists matrices  $X_1 \geq 0$ ,  $X_2 \geq 0$ ,  $X_3 \geq 0$ ,  $K_i$ ,  $L_i^1$  and  $L_i^2$  and scalars  $\overline{\gamma} \rho$  and  $\tau$  such that, for i = 1, 2, ..., r

$$\begin{split} & \Phi^{1,1} = \rho I + Q_1 - X_1 & \Phi^{2,2} = \rho I + Q_2 - X_2 & \Phi^{3,3} = \rho I + Q_3 - X_3 & \Phi^{5,5} = -\rho \varepsilon I + \tau^{-1} \delta I \\ & \Phi^{6,4}_i = X_1 (B_i - G_i) & \Phi^{7,2}_{ij} = X_2 A_i - L_j^1 C_i & \Phi^{7,3}_{ij} = X_2 G_i - L_j^1 W_i & \Phi^{8,3}_{ij} = X_3 - L_j^2 W_i \end{split}$$

The observer gains and the attenuation level are obtained by :

$$H_i^1 = X_2^{-1} L_i^1$$
,  $H_i^2 = X_3^{-1} L_i^2$  and  $\gamma = \sqrt{\bar{\gamma}}$ 



#### Takagi-Sugeno model

$$\begin{cases} x(k+1) = \sum_{i=1}^{2} \mu_{i}(u(k)) (A_{i}x_{f}(k) + B_{i}u_{f}(k) + G_{i}f(k)) \\ y(k) = \sum_{i=1}^{2} \mu_{i}(u(k)) (C_{i}x_{f}(k) + D_{i}u_{f}(k) + W_{i}f(k)) \end{cases}$$

with

$$A_{1} = \begin{pmatrix} -0.5 & 0.1 \\ -1 & -1 \end{pmatrix} \quad A_{2} = \begin{pmatrix} 0 & 0.2 \\ -0.45 & -0.7 \end{pmatrix} \quad B_{1} = \begin{pmatrix} 0.4 \\ 0.5 \end{pmatrix} \quad B_{2} = \begin{pmatrix} 0.6 \\ 0.4 \end{pmatrix}$$

$$G_{1} = \begin{pmatrix} 0.2 \\ 0.4 \end{pmatrix} \quad G_{2} = \begin{pmatrix} 0.5 \\ 0.5 \end{pmatrix}$$

$$C_{1} = \begin{pmatrix} 0.2 & 0 \end{pmatrix} \quad C_{2} = \begin{pmatrix} 0.4 & 0.1 \end{pmatrix} \quad W_{1} = -0.3 \quad W_{2} - 0.4$$

$$\mu_{1}(u(k)) = \frac{1 - \tanh(0.5 - u(k))}{2} \quad \mu_{2}(u(k)) = \frac{1 + \tanh(0.5 - u(k))}{2}$$

The nominal input signal is :  $u(k) = 0.5\cos(\sin(0.1k)0.1k)$ .

The FT Controller is designed for :  $\alpha_0 = 0.1$  and  $\lambda = 1.3$ 

The fault affecting the system is:  $f(k) = e^{0.5k-10}$ , for  $9 \le k \le 17$ 

### Simulation results – state and fault estimation



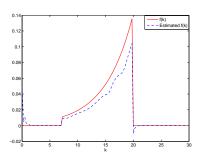


FIGURE: Fault and its estimation

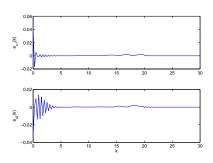


FIGURE: State estimation errors

# Simulation results - trajectory tracking



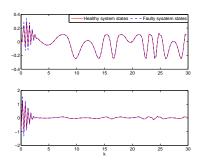


FIGURE: Reference model states vs. faulty system ones with FTC

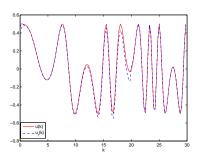


FIGURE: Nominal and FTC control inputs



### Conclusions

Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.



### Conclusions

- Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.
- State and fault estimation are achieved simultaneously



### Conclusions

- Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.
- State and fault estimation are achieved simultaneously
- ► Fault tolerant control with reference trajectory tracking



#### Conclusions

- Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.
- State and fault estimation are achieved simultaneously
- ► Fault tolerant control with reference trajectory tracking
- The problem of FTC design is expressed via an optimization problem subject to LMI (Linear Matrix Inequality) constraints.



#### Conclusions

- Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.
- State and fault estimation are achieved simultaneously
- Fault tolerant control with reference trajectory tracking
- The problem of FTC design is expressed via an optimization problem subject to LMI (Linear Matrix Inequality) constraints.
- A similar solution has been proposed for FTC of continuous time Takagi-Sugeno systems (MED'2011)



#### Conclusions

- Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.
- State and fault estimation are achieved simultaneously
- Fault tolerant control with reference trajectory tracking
- The problem of FTC design is expressed via an optimization problem subject to LMI (Linear Matrix Inequality) constraints.
- A similar solution has been proposed for FTC of continuous time Takagi-Sugeno systems (MED'2011)

### Perspectives

▶ Study of the unmeasurable premise variable case  $(\xi(t) = x(t))$ .



#### Conclusions

- Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.
- State and fault estimation are achieved simultaneously
- Fault tolerant control with reference trajectory tracking
- The problem of FTC design is expressed via an optimization problem subject to LMI (Linear Matrix Inequality) constraints.
- A similar solution has been proposed for FTC of continuous time Takagi-Sugeno systems (MED'2011)

- ▶ Study of the unmeasurable premise variable case  $(\xi(t) = x(t))$ .
- Comparison with multiple integral observer approach



#### Conclusions

- Active fault tolerant control law for nonlinear systems represented by a Takagi-Sugeno structure.
- State and fault estimation are achieved simultaneously
- Fault tolerant control with reference trajectory tracking
- The problem of FTC design is expressed via an optimization problem subject to LMI (Linear Matrix Inequality) constraints.
- A similar solution has been proposed for FTC of continuous time Takagi-Sugeno systems (MED'2011)

- ▶ Study of the unmeasurable premise variable case  $(\xi(t) = x(t))$ .
- Comparison with multiple integral observer approach
- Implementation of a bank of different controller each of them dedicated to a particular kind of fault and design of a switching control law depending on the measured performances.